

Features

- V_{IN} Input Range from 7.5V to 36V
- Accumulative Cell Voltage Monitor: 8-to-1 Analog Multiplexer with divided ratio accuracy: $1/n \pm 0.5\%$
- Reverse-current prevention switching scan frequency of 100Hz
- Requires fewer MCU A/D Converters (ADCs)
- 5V/30mA internal Voltage Regulator with $\pm 1\%$ accuracy
- Operating Temperature Range: -40°C to $+85^{\circ}\text{C}$
- Package Type: 16-pin NSOP

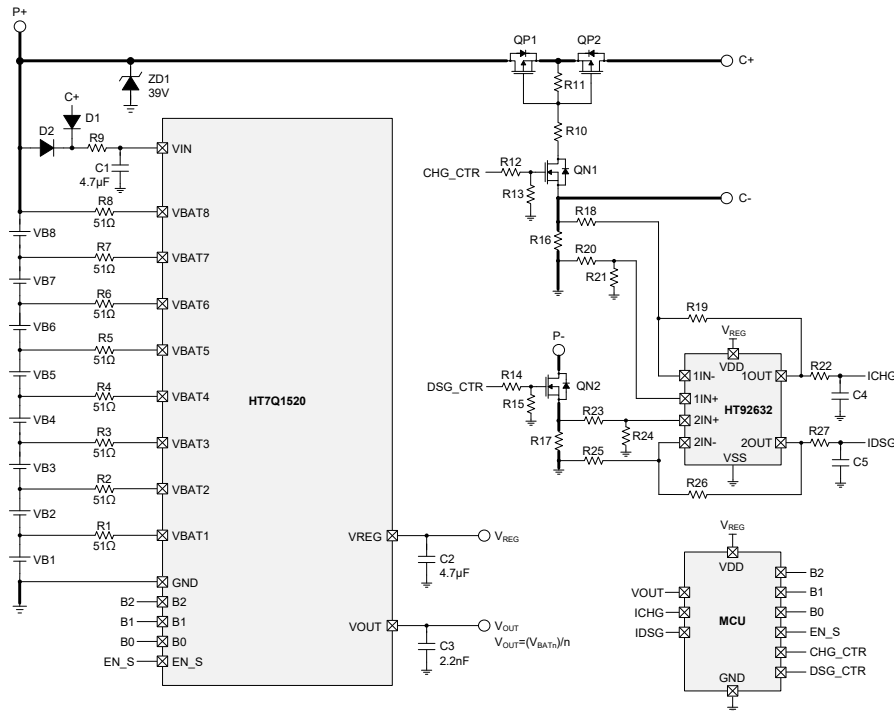
Applications

- Electric power tools
- Handheld vacuum cleaners

General Description

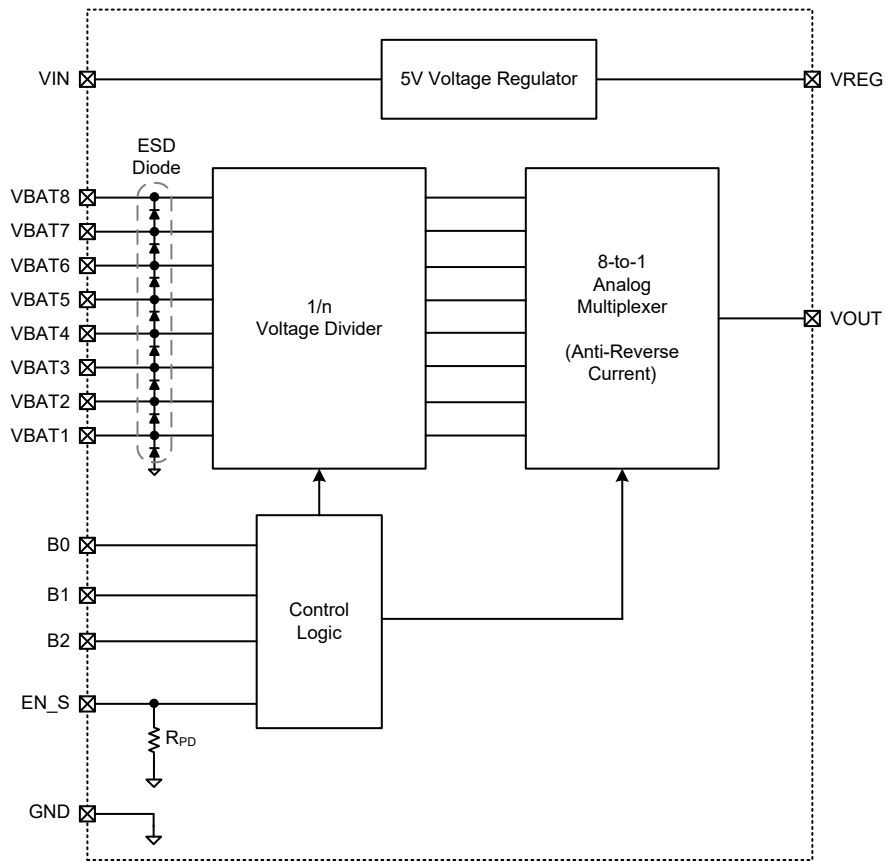
The HT7Q1520 is a high voltage analog-front-end IC for 3 to 8 cell Li-ion rechargeable battery protection. It consists of an accumulative cell voltage monitor and a high accuracy voltage regulator. The device is designed to monitor an accumulative voltage from 1 to N and outputs the divide-by-N voltage to the analog multiplexer with a $\pm 0.5\%$ divided ratio accuracy. The anti-reverse current switch is implemented to prevent a backflow current even if V_{OUT} is higher than V_{BATn} . Each divided accumulative cell voltage can be observed sequentially on pin $VBATn$ which will benefit MCUs with a lower number of ADCs. There are 3 control bits, B0, B1 and B2, to select which terminal voltage outputs with a maximum 100Hz scanning frequency when $C_{OUT}=2.2\text{nF}$. The enable pin, EN_S , is used to close all switches and the output voltage is pulled down by an internal $1\text{M}\Omega$ resistor. An integrated regulator provides a 5V supply to the MCU with a 30mA driving current capability and which has $\pm 1\%$ precision. The voltage regulator is always active even if the EN_S pin is cleared to a logic low level.

Typical Application Circuit

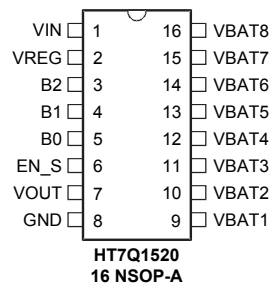


- Note: 1. If less than 8 serial batteries are used, connect the unused $VBATn$ to the highest voltage potential. Do not leave any $VBATn$ pin floating to prevent damage to the device.
2. The bold lines indicate that these connections need to be as short as possible.
3. VIN pin should not be floating to prevent abnormal operations.

Functional Block Diagram



Pin Assignment



Pin Description

Pin No.	Pin Name	Type	Pin Discription
1	VIN	P	Input supply voltage. Connect to the top VBATn
2	VREG	O	Regulator 5V/30mA output. Connect a 4.7 μ F capacitor to GND
3	B2	I	8-to-1 analog multiplexer selection bit - MSB
4	B1	I	8-to-1 analog multiplexer selection bit
5	B0	I	8-to-1 analog multiplexer selection bit - LSB
6	EN_S	I	Enable terminal for the 8-to-1 analog multiplexer. Connected to 1M Ω internal pull low resistor
7	VOUT	O	8-to-1 analog multiplexer output. Connect a 2.2nF capacitor to GND.
8	GND	G	Ground terminal
9	VBAT1	I	Battery cell 1 positive terminal and battery cell 2 negative terminal
10	VBAT2	I	Battery cell 2 positive terminal and battery cell 3 negative terminal
11	VBAT3	I	Battery cell 3 positive terminal and battery cell 4 negative terminal
12	VBAT4	I	Battery cell 4 positive terminal and battery cell 5 negative terminal
13	VBAT5	I	Battery cell 5 positive terminal and battery cell 6 negative terminal
14	VBAT6	I	Battery cell 6 positive terminal and battery cell 7 negative terminal
15	VBAT7	I	Battery cell 7 positive terminal and battery cell 8 negative terminal
16	VBAT8	I	Battery cell 8 positive terminal

Absolute Maximum Ratings

Parameter		Value	Unit
V _{IN}		-0.3 to +40	V
V _{REG} , V _{OUT} , B2, B1, B0, EN_S, V _{BAT1}		-0.3 to +5.5	V
$\Delta[V_{BATi} - V_{BAT(i-1)}]$, i=8, 7, 6, 5, 4, 3, 2		-0.3 to +5.5	V
Operating Temperature Range		-40 to +85	°C
Maximum Junction Temperature		+125	°C
Storage Temperature Range		-65 to +160	°C
Lead Temperature (Soldering 10sec)		+260	°C
ESD Susceptibility	Human Body Model	\pm 6000	V
	Machine Model	\pm 350	V
Junction-to-Ambient Thermal Resistance, θ_{JA}	16NSOP (150mil)	100	°C/W

Recommended Operating Ratings

Parameter	Value	Unit
V _{IN}	7.5 to 36.0	V
T _a	-40 to +85	°C

Note that Absolute Maximum Ratings indicate limitations beyond which damage to the device may occur. Recommended Operating Ratings indicate conditions for which the device is intended to be functional, but do not guarantee specified performance limits.

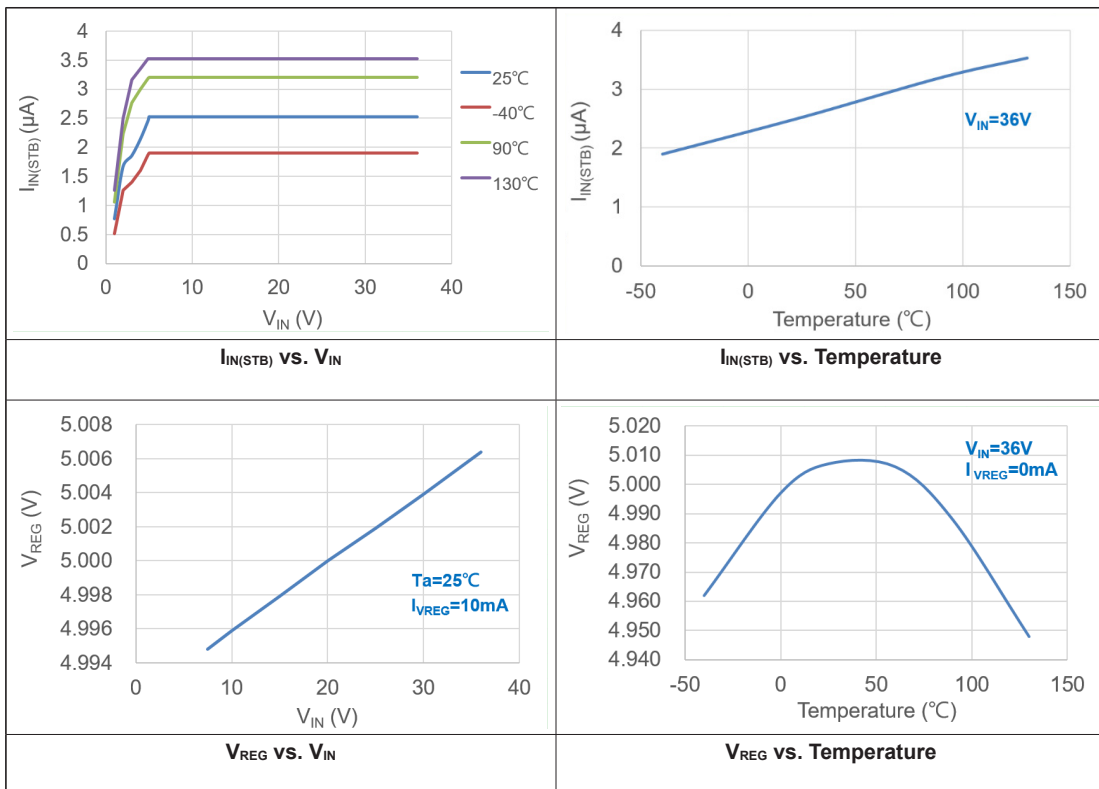
Electrical Characteristics

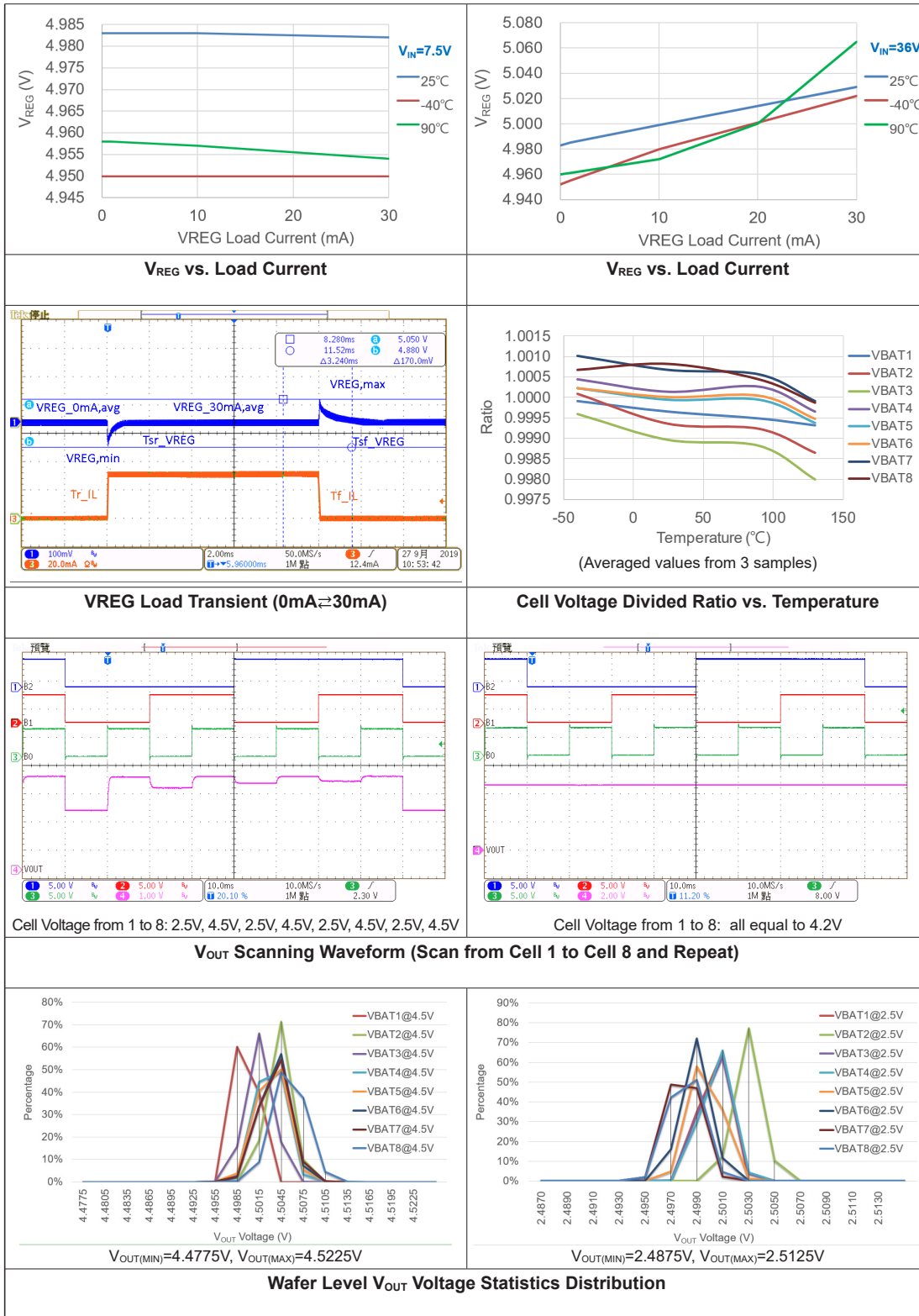
 $V_{IN}=36V$, $C_{REG}=4.7\mu F$, $C_{OUT}=2.2nF$ and $T_a=25^\circ C$, unless otherwise specified.

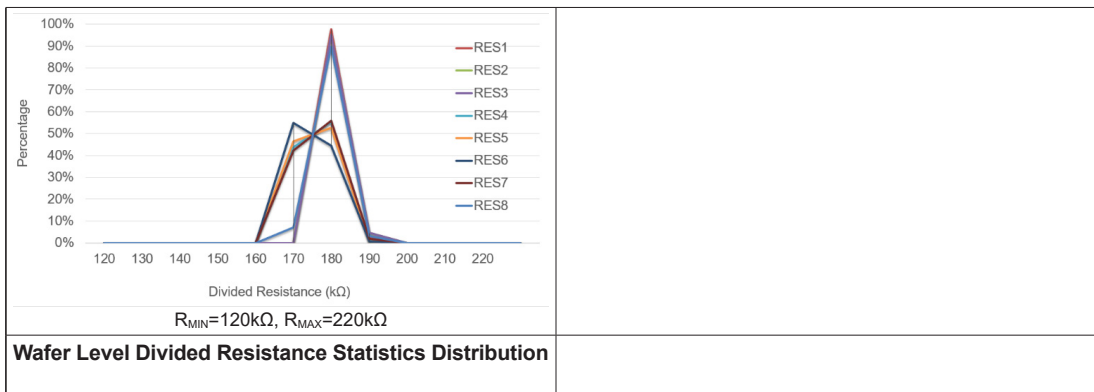
Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
Supply and Input						
V_{IN}	Supply Voltage	—	7.5	—	36.0	V
$I_{IN(SCAN)}$	Supply Current – Scan	$EN_S=1$, Scan B2~B0 with frequency 100Hz	—	2.5	4.0	μA
$I_{IN(STB)}$	Supply Current – Standby	$B0=B1=B2=EN_S=0$	—	2.5	4.0	μA
V_{Bi}	Cell Voltage Range	$i=1\sim 8$	2.5	—	4.5	V
I_{Bi}	Cell Input Leakage Current	$V_{BATi}=5V \times i$, $EN_S=0$, $V_{IN}=V_{BATi}$, $i=1\sim 8$	-0.1	—	0.1	μA
$I_{Bi(ACT)}$	Cell Input Current when Monitoring Voltage	$V_{BATi}=4.2V \times i$, $EN_S=1$, $V_{IN}=36V$, $i=1\sim 8$	19	24	35	μA
$I_{B1(REV)}$	VBAT1 Input Reverse Current	$V_{BAT8}=36V$, $V_{BAT1}=2.5V$, B2~B0=0b'111, Measure I_{B1}	-0.1	—	0.1	μA
Voltage Regulator						
V_{REG}	Regulator Output Voltage	$I_{LOAD}=10mA$	4.95	5.00	5.05	V
I_{REG}	Regulator Output Current	$V_{IN}=7.5V$, $T_a=-40\sim 85^\circ C$	30	—	—	mA
ΔV_{REG}	Load Regulation	$I_{LOAD}=0\sim 30mA$	—	50	—	mV
$\Delta V_{REG}/(V_{REG} \times \Delta V_{IN})$	Line Regulation	$V_{IN}=7.5V\sim 36V$, $I_{LOAD}=10mA$	—	0.02	—	%/V
$\Delta V_{REG}/(V_{REG} \times \Delta T_a)$	Temperature Coefficient	$I_{LOAD}=1mA$, $T_a=-40\sim 85^\circ C$	—	± 100	—	ppm/ $^\circ C$
Accumulative Cell Voltage Monitor						
R	Divided Resistance	—	120	170	220	k Ω
Ratio8(NORM)	VBAT8 Accumulative Cell Voltage Divided Ratio (Normalised)	$V_{BAT8}=20V\sim 36V$, B2~B0=0b'111, $EN_S=1$	0.995	1.000	1.005	—
		$V_{BAT8}=20V\sim 36V$, B2~B0=0b'111, $EN_S=1$, $T_a=-40\sim 85^\circ C$	0.990	1.000	1.010	—
Ratio7(NORM)	VBAT7 Accumulative Cell Voltage Divided Ratio (Normalised)	$V_{BAT7}=17.5V\sim 31.5V$, B2~B0=0b'110, $EN_S=1$	0.995	1.000	1.005	—
		$V_{BAT7}=17.5V\sim 31.5V$, B2~B0=0b'110, $EN_S=1$, $T_a=-40\sim 85^\circ C$	0.990	1.000	1.010	—
Ratio6(NORM)	VBAT6 Accumulative Cell Voltage Divided Ratio (Normalised)	$V_{BAT6}=15V\sim 27V$, B2~B0=0b'101, $EN_S=1$	0.995	1.000	1.005	—
		$V_{BAT6}=15V\sim 27V$, B2~B0=0b'101, $EN_S=1$, $T_a=-40\sim 85^\circ C$	0.990	1.000	1.010	—
Ratio5(NORM)	VBAT5 Accumulative Cell Voltage Divided Ratio (Normalised)	$V_{BAT5}=12.5V\sim 22.5V$, B2~B0=0b'100, $EN_S=1$	0.995	1.000	1.005	—
		$V_{BAT5}=12.5V\sim 22.5V$, B2~B0=0b'100, $EN_S=1$, $T_a=-40\sim 85^\circ C$	0.990	1.000	1.010	—
Ratio4(NORM)	VBAT4 Accumulative Cell Voltage Divided Ratio (Normalised)	$V_{BAT4}=10V\sim 18V$, B2~B0=0b'011, $EN_S=1$	0.995	1.000	1.005	—
		$V_{BAT4}=10V\sim 18V$, B2~B0=0b'011, $EN_S=1$, $T_a=-40\sim 85^\circ C$	0.990	1.000	1.010	—
Ratio3(NORM)	VBAT3 Accumulative Cell Voltage Divided Ratio (Normalised)	$V_{BAT3}=7.5V\sim 13.5V$, B2~B0=0b'010, $EN_S=1$	0.995	1.000	1.005	—
		$V_{BAT3}=7.5V\sim 13.5V$, B2~B0=0b'010, $EN_S=1$, $T_a=-40\sim 85^\circ C$	0.990	1.000	1.010	—

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
Ratio2 _(NORM)	VBAT2 Accumulative Cell Voltage Divided Ratio (Normalised)	V _{BAT2} =5V~9V, B2~B0=0b'001, EN_S=1	0.995	1.000	1.005	—
		V _{BAT2} =5V~9V, B2~B0=0b'001, EN_S=1, Ta=-40~85°C	0.990	1.000	1.010	—
Ratio1 _(NORM)	VBAT1 Accumulative Cell Voltage Divided Ratio (Normalised)	V _{BAT1} =2.5V~4.5V, B2~B0=0b'000, EN_S=1	0.995	1.000	1.005	—
		V _{BAT1} =2.5V~4.5V, B2~B0=0b'000, EN_S=1, Ta=-40~85°C	0.990	1.000	1.010	—
Input Logic						
f _{MAX}	B2~B0 Maximum Scan Frequency	C _{OUT} = 2.2nF	100	—	—	Hz
V _{IH}	Input Logic High Threshold	V _{IN} =7.5~36V, B2, B1, B0 and EN_S pins	2.5	—	—	V
V _{IL}	Input Logic Low Threshold	V _{IN} =7.5~36V, B2, B1, B0 and EN_S pins	—	—	0.8	V
R _{PD}	Pull Low Resistance	EN_S pin	—	1	—	MΩ
I _{LEAK}	Input Logic Leakage	V _{IN} =7.5V~36V, B2, B1 and B0 pins	—	—	0.1	μA

Typical Performance Characteristics







Functional Description

Accumulative Cell Voltage Monitor

The accumulative cell voltage monitor consists of high voltage switches, voltage dividers and an 8-to-1 analog multiplexer as shown in Fig 1. The high voltage switches are implemented using an anti-reverse current topology which provides isolation between the output voltage and the unselected VBATs.

B2, B1 and B0 are used to control the p-type switches S1~S8 only if EN_S='H'. The control truth table is shown below. This produces an accumulative cell voltage, V_{BATn}, divided by “n” on V_{OUT}. This accurate ±0.5% voltage divided ratio is designed to minimise any mismatch errors.

EN_S	B2	B1	B0	S8	S7	S6	S5	S4	S3	S2	S1	V _{OUT} (V)
0	X	X	X	0	0	0	0	0	0	0	0	0
1	0	0	0	0	0	0	0	0	0	0	1	V _{BAT1} ×1/1
1	0	0	1	0	0	0	0	0	0	1	0	V _{BAT2} ×1/2
1	0	1	0	0	0	0	0	0	1	0	0	V _{BAT3} ×1/3
1	0	1	1	0	0	0	0	1	0	0	0	V _{BAT4} ×1/4
1	1	0	0	0	0	0	1	0	0	0	0	V _{BAT5} ×1/5
1	1	0	1	0	0	1	0	0	0	0	0	V _{BAT6} ×1/6
1	1	1	0	0	1	0	0	0	0	0	0	V _{BAT7} ×1/7
1	1	1	1	1	0	0	0	0	0	0	0	V _{BAT8} ×1/8
Power on states				0	0	0	0	0	0	0	0	0

Table 1: Accumulative Cell Voltage Monitor Truth Table

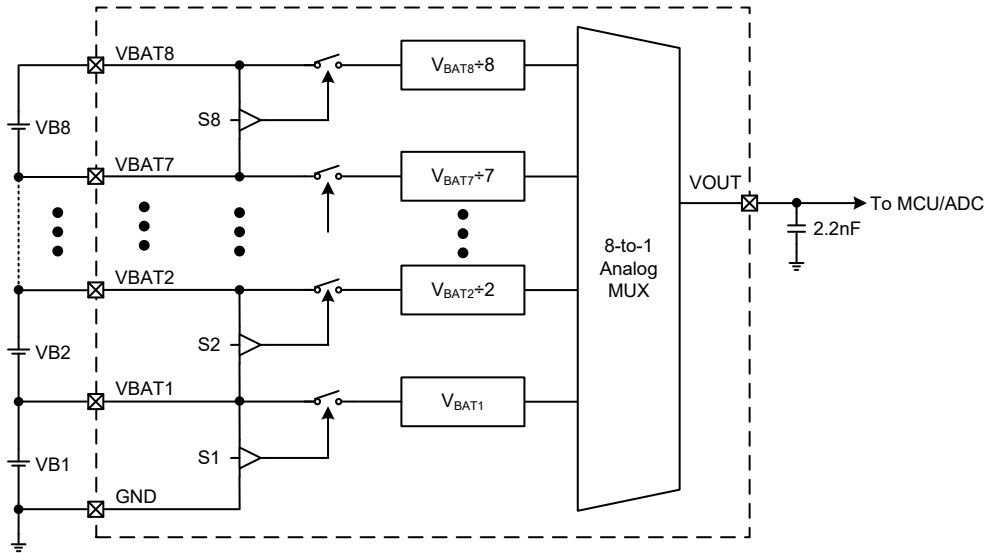
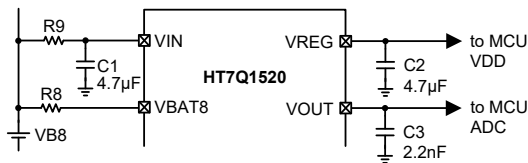


Fig 1. Accumulative Cell Voltage Monitor Functional Block

Application Information

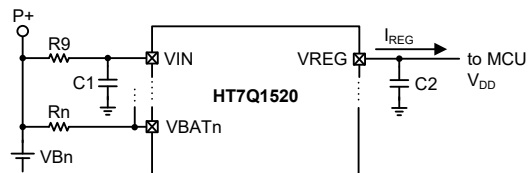
VIN, VREG, VOUT Capacitors

The VIN input capacitor C_{IN} and VREG output capacitor C_{REG} are recommended to have a value of $4.7\mu F$ respectively and for better input noise filtering and output load transient behavior. A recommended $2.2nF$ noise filtering capacitor should be connected between VOUT and GND terminals. Note that higher noise capacitance value of C_{OUT} will lower the acceptable scan frequency.



VIN and VBATn Spike Suppression Resistors

The appropriate VIN and VBATn spike suppression resistors corresponded to R_9 and R_n , lower the voltage spike and inrush current applied on any I/O pins, which can improve the stability of V_{REG} that provides the power source to the external MCUs. Especially when there are spike filter capacitors connected from P+ or C+ to GND, the VBATn spike suppression resistors (R_n) are necessary. The recommended resistance values of VIN and VBATn resistors depend on the number of battery cells and are listed in Table 2.



Battery Cells	VIN Recommended Resistor (R9)	VBATn Recommended Resistor (Rn)
3S	18Ω	100Ω
4S	100Ω	51Ω
5S	180Ω	51Ω
6S	270Ω	51Ω
7S	360Ω	51Ω
8S	430Ω	51Ω

Table 2: Spike Suppression Recommended Resistor

It is necessary to select an appropriate package for VIN spike suppression resistor (R_9) in order to prevent it being damaged from overheated. The maximum power on R_9 can be easily calculated by the following formula:

$$P_{R9,MAX} = (I_{REG})^2 \times R_9$$

Where, the I_{REG} is the maximum VREG output current. The R_9 is the resistance value of VIN spike suppression resistor.

It is recommended to select the resistor package that its maximum rated power is greater than twice the $P_{R9,MAX}$.

Cell Voltage Monitor Scan Frequency

The HT7Q1520 device can output accumulative cell voltage to external MCU/ADC for monitoring battery voltage status. The Fig2 and Fig3 show the timing diagrams of cell voltage monitor scanning for 5S and 8S applications. The HT7Q1520 starts to charge the VOUT capacitor from 0V to the selected cell voltage when the EN_S receives an 'L' to 'H' signal. In order to insure

the external MCU A/D conversion accuracy, the A/D conversion procedure has to wait before the HT7Q1520's VOUT capacitor is fully charged. The suggested minimum waiting time is 5ms after EN_S rises from 'L' to 'H' or cell voltage monitor channel is switched. It is recommended that the maximum scan frequency for accumulative cell voltage monitoring is less than 100Hz and that EN_S='0' when the voltage scanning procedure has finished for power saving purposes.

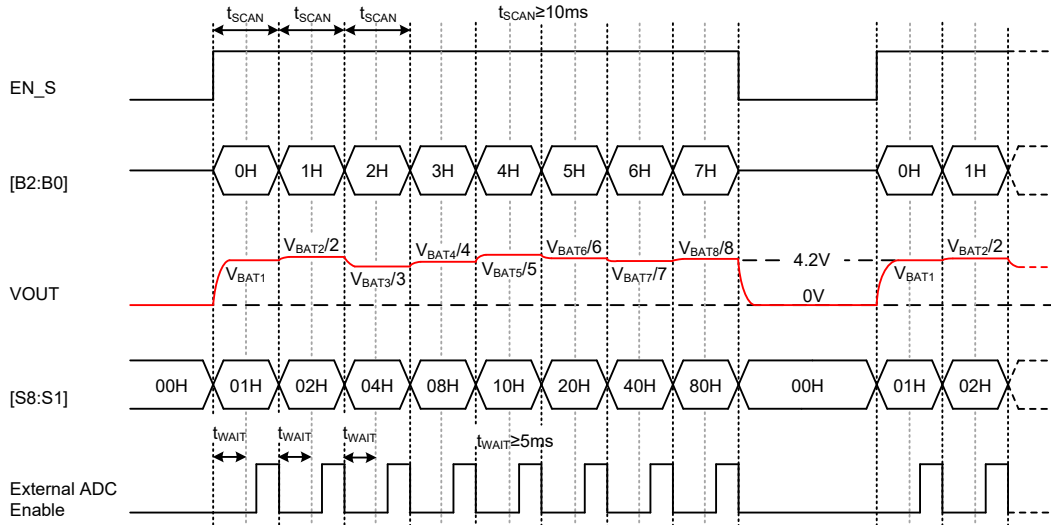


Fig 2. 8S Battery Monitoring Timing

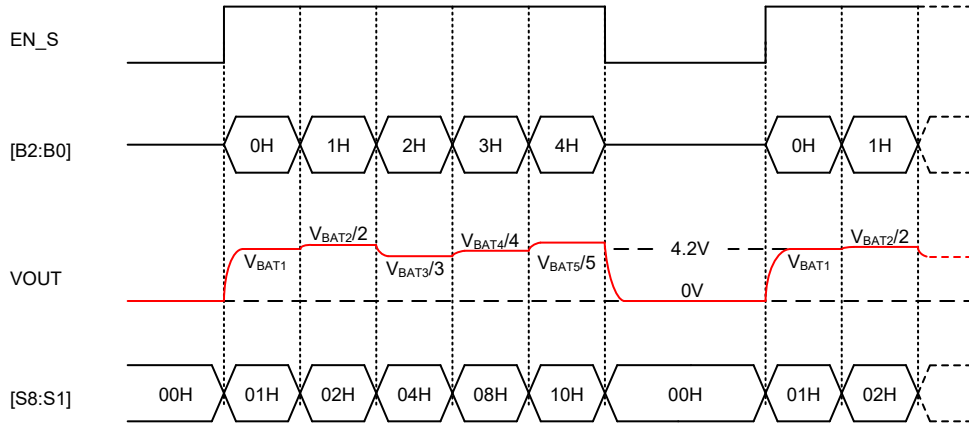


Fig 3. 5S Battery Monitoring Timing

Acquiring Cell Voltage Monitor Output with External MCU ADC

The HT7Q1520 could output the battery cell voltage to an external MCU for monitoring the battery status when it is in charging or discharging state. As shown in Fig.4, the external MCU ADC samples the V_{OUT} voltage via an ADC sampling capacitor (C_{SAMPLE}), which is typically 5pF to 50pF. Due to the charge sharing effect, the voltage on the V_{OUT} capacitor

(C_{OUT}) drops while the ADC samples that with the initially zero-voltage C_{SAMPLE} . A voltage-drop occurs on the C_{OUT} after the first ADC sampling, and then it needs a recharge time to recharge the C_{OUT} . Referring to Fig.5, a pre-charge procedure of C_{SAMPLE} during the V_{OUT} capacitor recharge time is recommended to minimize the charge sharing effect by means of performing several sampling without clearing the charges in the C_{SAMPLE} .

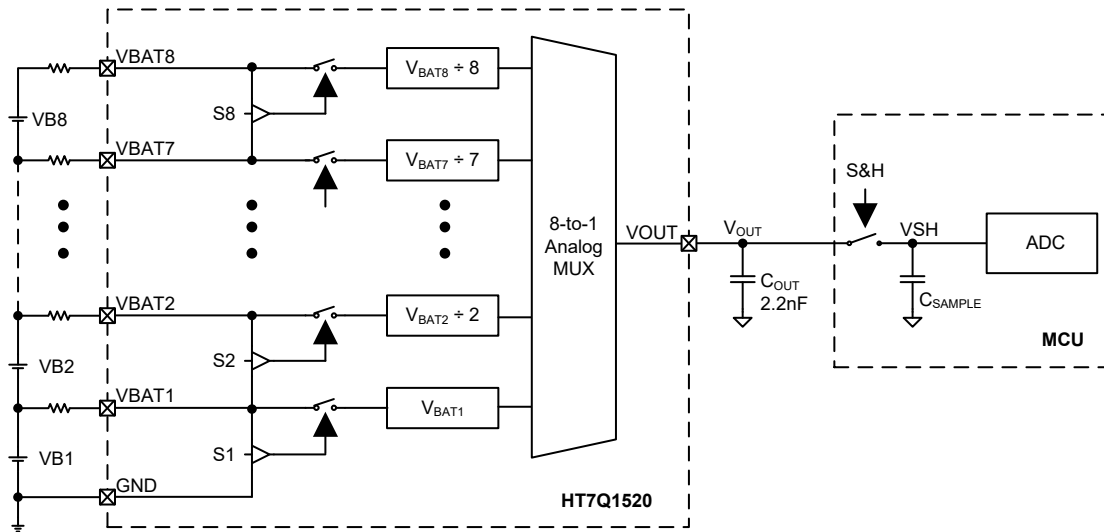


Fig 4. HT7Q1520 + MCU/ADC Functional Block

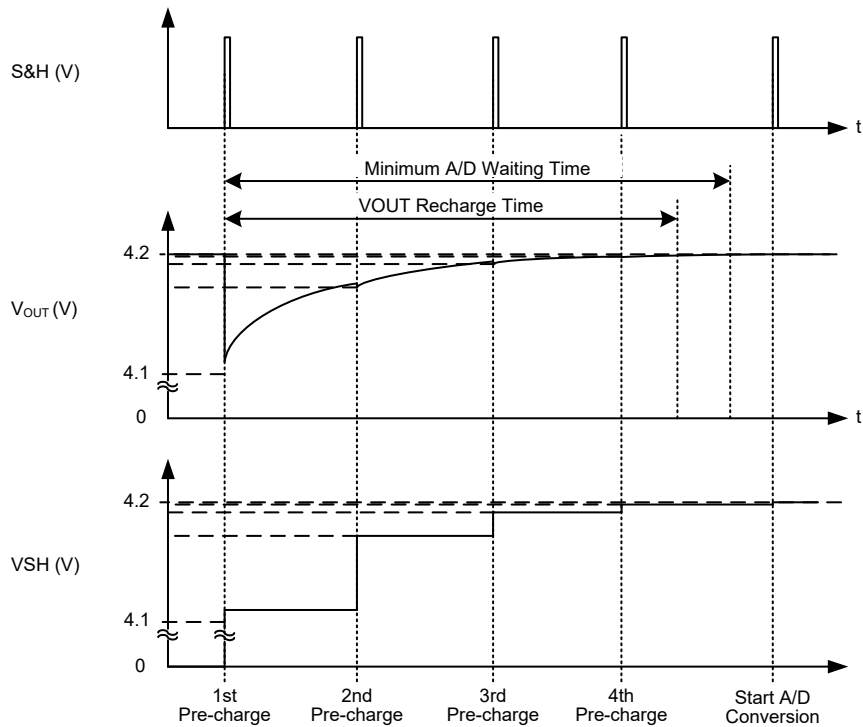


Fig 5. ADC Sampling Recommended Timing

For the best conversion accuracy, the A/D conversion has to wait before the C_{OUT} recharge is complete. The recommended minimum waiting time from the first sampling to A/D conversion beginning is listed in Table 3.

MCU/ADC Sampling Capacitance (pF)	5	10	20	30	40	50
Recommended minimum A/D waiting time (μ s)	370	660	950	1120	1240	1340

Table 3. ADC Recommended Waiting Time

Voltage Spike Suppression Method

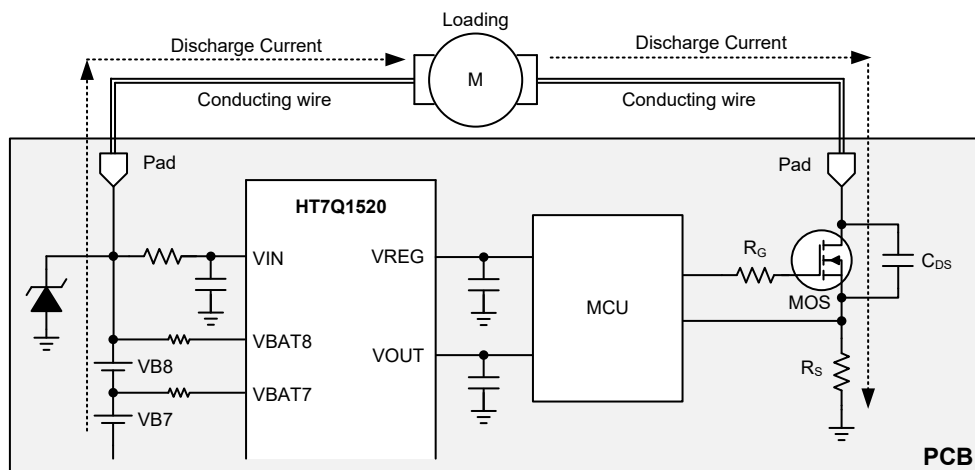


Fig 6. Simplified Typical BMS System Discharge Path Diagram

Most battery-management systems would monitor charge and discharge current to prevent over-current damage. Due to the parasitic inductance on conducting wires and PCB layout connections, a large voltage spike may occur while the MCU-controlled MOS rapidly shuts down the charge or discharge current, and this spike may damage HT7Q1520 VBATn or VIN pins. Any voltage spike on the VBATn and VIN pins should not over the limitation in Absolute Maximum Ratings, which is 40V. Four recommended measures listed below would help to reduce the voltage spike.

1. Ensure that the external conducting wires and PCB layout connections where large charge or discharge current flows are as short as possible.
2. Adjust the slew rate of MOS switch with the Gate resistor R_G . Turn off the MOS switch with a slower slew rate can lower voltage spike, however the trade off is a slower protection response time.
3. Add a capacitor (C_{DS}) between the drain and source nodes of the MOS switch as shown in Fig6. The recommended capacitance is 0.1μ F to 0.22μ F.
4. Add a 39V Zener diode between the highest voltage potential node of the battery cells and GND.

PCB Layout Considerations

The following placement/layout guidelines are suggested for the sake of noise reduction and voltage spike suppression.

1. The VIN filter capacitor and resistor must be placed close to the VIN pin. [Region (A)]
2. The VREG regulation and noise filter capacitor must be placed close to the VREG pin. [Region (B)]
3. The VOUT noise filter capacitor must be placed close to the VOUT pin. [Region (C)]
4. Placing a ground line between VIN and VREG lines would improve the ability of resisting surge.
5. The tracks where large current would flow through should be wide and short to suppress the voltage spike at the time when MOS switch changes its ON/OFF state.
6. The power ground and signal ground should be connected at final output pad (B-) for less noise disturbance.

Evaluation Board

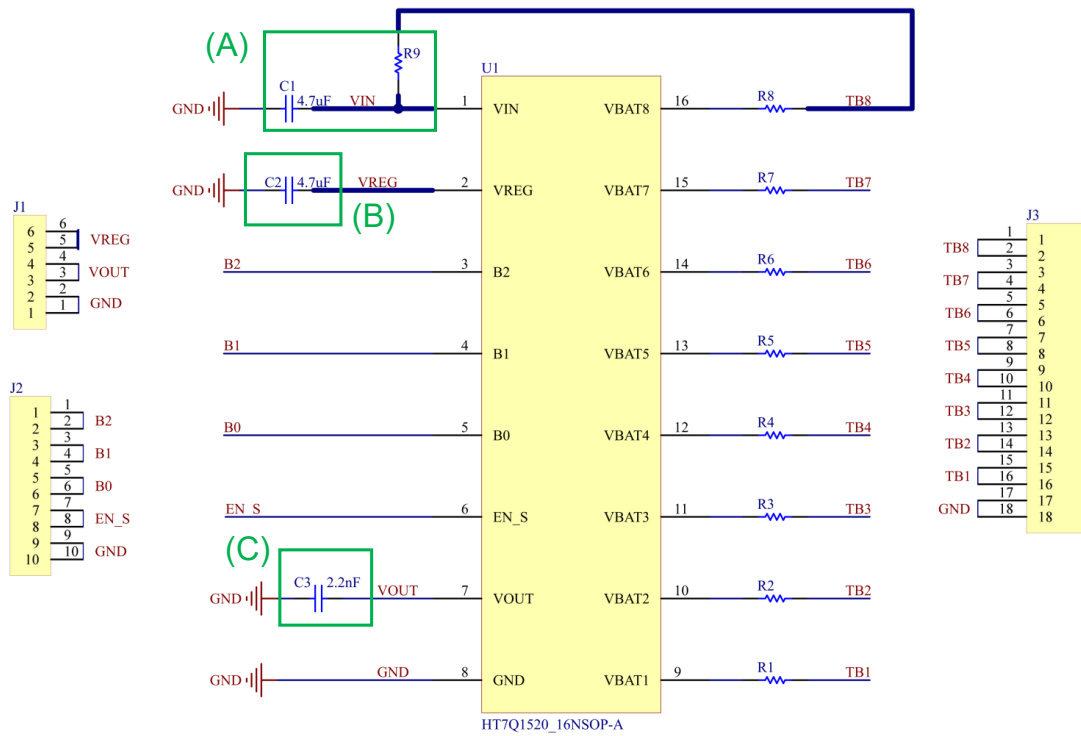


Fig 7. HT7Q1520 Evaluation Board Schematic

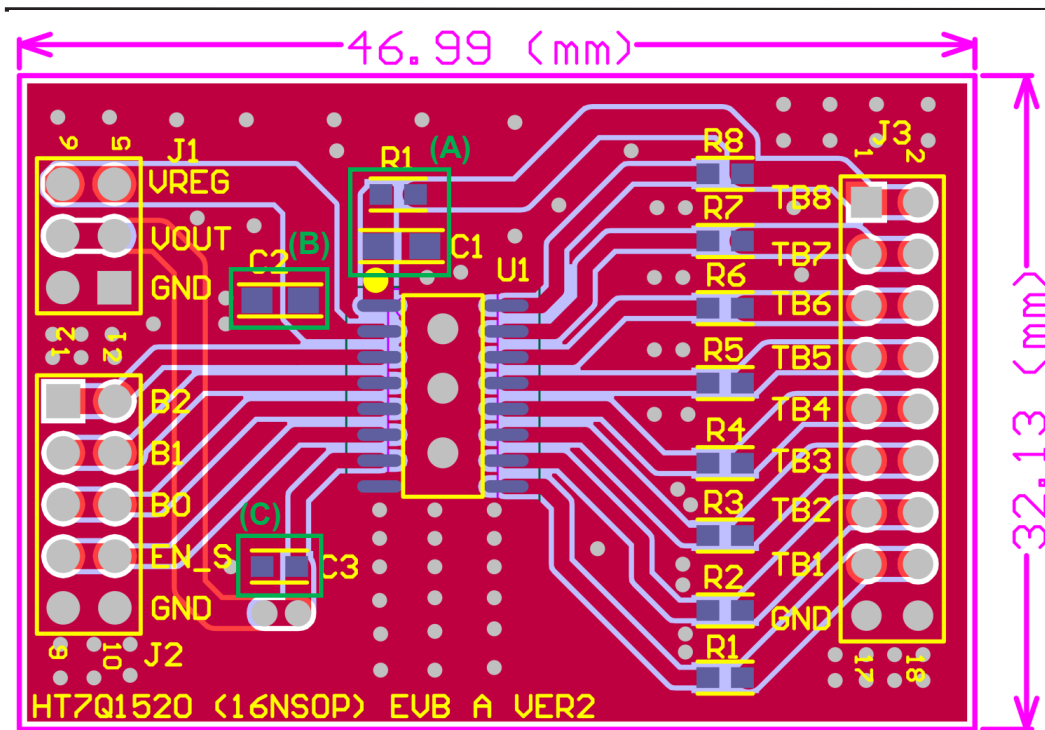


Fig 8. HT7Q1520 Evaluation Board Layout

Thermal Consideration

The maximum power dissipation depends upon the thermal resistance of the IC package, PCB layout, rate of surrounding airflow and difference between the junction and ambient temperature. The maximum power dissipation can be calculated using the following formula:

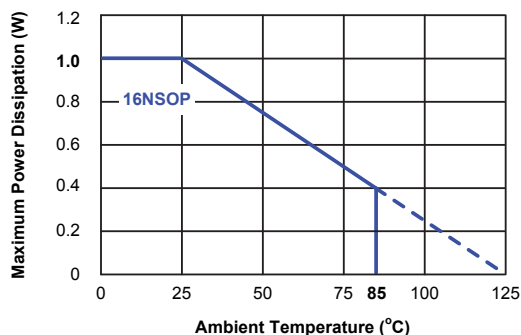
$$P_{D(MAX)} = (T_{J(MAX)} - T_a) / \theta_{JA} \quad (W)$$

where $T_{J(MAX)}$ is the maximum junction temperature, T_a is the ambient temperature and θ_{JA} is the junction-to-ambient thermal resistance of the IC package.

For maximum operating rating conditions, the maximum junction temperature is 125°C. However, it is recommended that the maximum junction temperature does not exceed 125°C during normal operation to maintain high reliability. The de-rating curve of the maximum power dissipation is shown below:

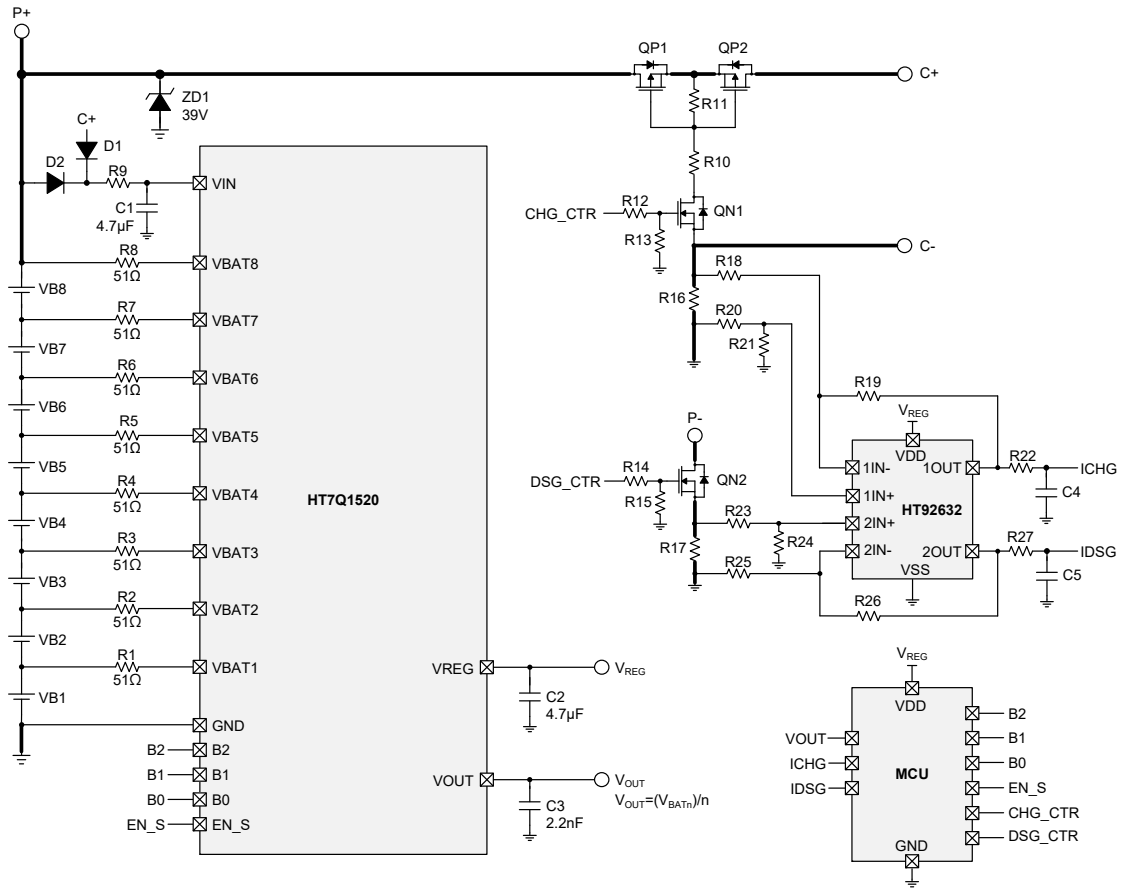
$$P_{D(MAX)} = (125^{\circ}C - 25^{\circ}C) / (100^{\circ}C/W) = 1.0W \quad (16NSOP)$$

For a fixed $T_{J(MAX)}$ of 125°C, the maximum power dissipation depends upon the operating ambient temperature and the package’s thermal resistance, θ_{JA} . The de-rating curve below shows the effect of rising ambient temperature on the maximum recommended power dissipation.

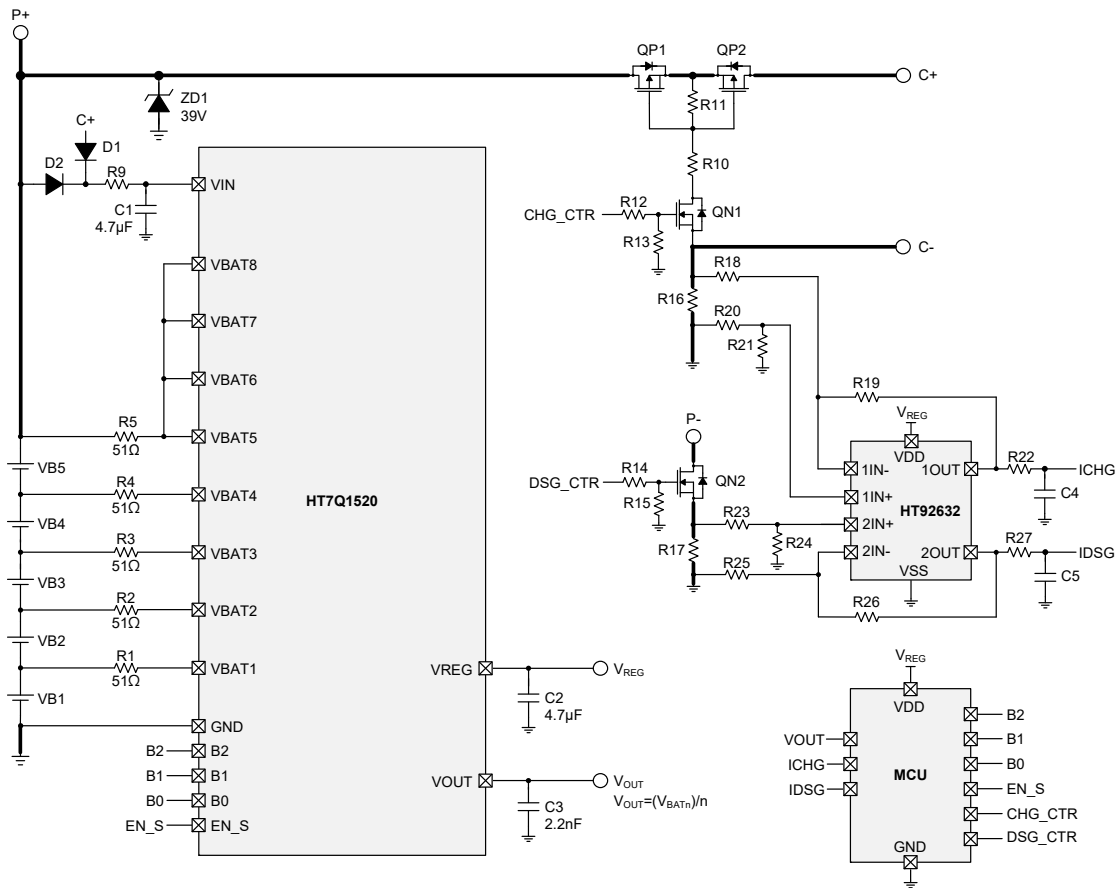


Application Circuits

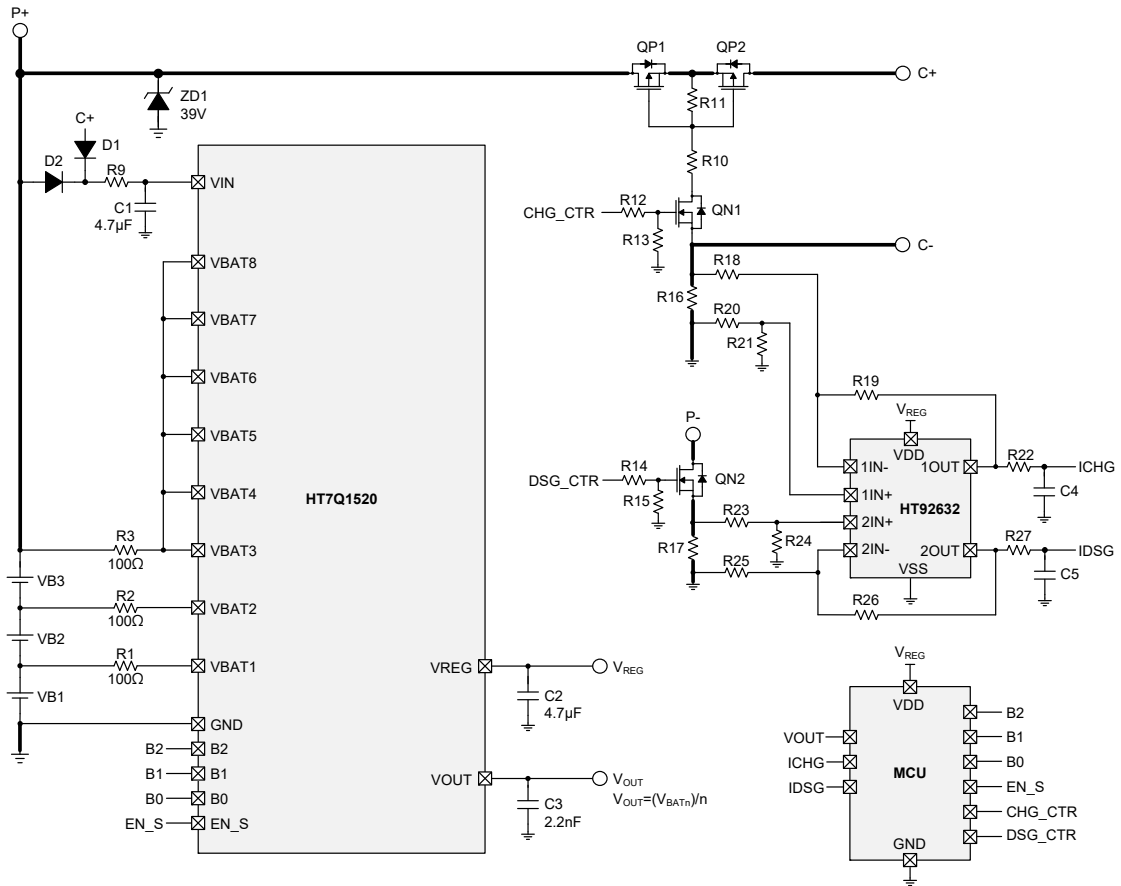
8S Battery Monitoring Typical Application Circuit



5S Battery Monitoring Typical Application Circuit



3S Battery Monitoring Typical Application Circuit



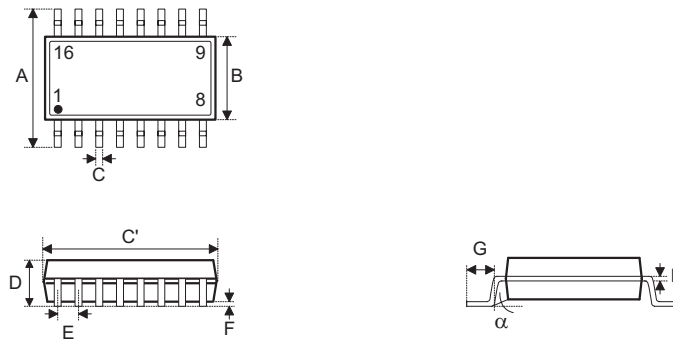
- Note: 1. If less than 8 serial batteries are used, connect the unused VBATn to the highest voltage potential. Do not leave any VBATn floating to prevent damage to the device.
2. For the selection of R9, R1, R2 and R3 resistances, refer to the “VIN and VBATn Spike Suppression Resistors” section for more details.
3. VIN pin should not be floating to prevent abnormal operations.

Package Information

Note that the package information provided here is for consultation purposes only. As this information may be updated at regular intervals users are reminded to consult the [Holtek website](#) for the latest version of the [package information](#).

Additional supplementary information with regard to packaging is listed below. Click on the relevant section to be transferred to the relevant website page.

- Further Package Information (include Outline Dimensions, Product Tape and Reel Specifications)
- Packing Materials Information
- Carton information

16-pin NSOP (150mil) Outline Dimensions


Symbol	Dimensions in inch		
	Min.	Nom.	Max.
A	—	0.236 BSC	—
B	—	0.154 BSC	—
C	0.012	—	0.020
C'	—	0.390 BSC	—
D	—	—	0.069
E	—	0.050 BSC	—
F	0.004	—	0.010
G	0.016	—	0.050
H	0.004	—	0.010
α	0°	—	8°

Symbol	Dimensions in mm		
	Min.	Nom.	Max.
A	—	6.000 BSC	—
B	—	3.900 BSC	—
C	0.31	—	0.51
C'	—	9.900 BSC	—
D	—	—	1.75
E	—	1.270 BSC	—
F	0.10	—	0.25
G	0.40	—	1.27
H	0.10	—	0.25
α	0°	—	8°

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